

7.0V to 9.5V

38mA(Typ)

0.002%(Typ)

2.3Vrms (Typ)

-100dB (Typ)

+15dB to -79dB

3.8µVrms(Typ)

Sound Processor with Built-in 3-band Equalizer BD37534FV

General Description

BD37534FV is a sound processor with built-in 3-band equalizer for car audio. A stereo input selector is available that functions to switch single end input and ground isolation input, input-gain control, main volume, loudness, 5ch fader volume, LPF for subwoofer and mixing input. Moreover, "Advanced switch circuit", which is an original ROHM technology, can reduce various switching noise (ex. No-signal, low frequency like 20Hz & large signal inputs). Also, "Advanced switch" makes microcomputer control easier, and constructs a high quality car audio system.

Features

- Reduced switching noise of input gain control, mute, main volume, fader volume, bass, middle, treble, loudness by using advanced switch circuit
- Built-in differential input selector that can make various combination of single-ended / differential input.
- Built-in ground isolation amplifier inputs, which is ideal for external stereo input.
- Built-in input gain controller reduces volume switching noise of a portable audio input.
- Decreased number of external components due to built-in 3-band equalizer filter, LPF for subwoofer and loudness filter. It is possible to freely control the Q, Gv, fo of the 3-band equalizer and fc of the LPF and Gv of the loudness through the I²C BUS control
- A gain adjustment quantity of ±20dB with a 1 dB step gain adjustment is possible for the bass, middle and treble.
- Equipped with terminals for the subwoofer outputs. Also, the audio signal outputs of the front, rear and subwoofer can be chosen using the I²C BUS control.
- Built-in mixing input and mixing attenuator.
- Energy-saving design resulting in low current consumption is achieved by utilizing the BiCMOS process. It has the advantage of quality over scaling down the power heat control of the internal regulators..
- Input pins and output pins are organized and separately laid out to keep the signal flow in one direction. This consequently simplifies the pattern layout of the set board and decreases the board dimensions.
- It is possible to control the I²C BUS with 3.3V / 5V.

Applications

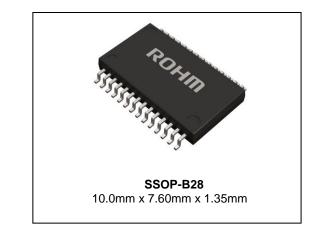
It is optimal for car audio systems. It can also be used for audio equipment of mini Compo, micro Compo, TV, etc.

Key Specifications

- Power Supply Voltage Range:
- Circuit Current (No Signal):
 Total Harmonic Distortion 1:
- Iotal Harmonic Distortion 1. (FRONT,REAR) 0.001%(Typ)
 Total Harmonic Distortion 2:
- (SUBWOOFER)
- Maximum Input Voltage:
- Cross-talk Between Selectors:
- Volume Control Range:
- Output Noise Voltage 1: (FRONT,REAR)
- Output Noise Voltage 2:
- (SUBWOOFER) 4.8µVrms(Typ) ■ Residual Output Noise Voltage: 1.8µVrms (Typ)
- Operating Temperature Range: -40°C to +85°C

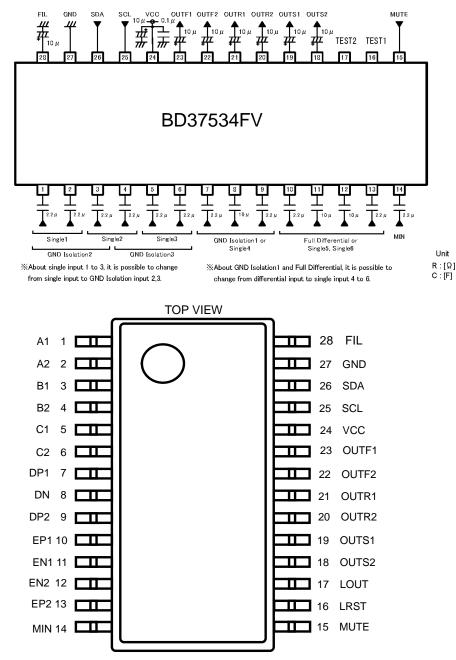
Package

W(Typ) x D(Typ) x H(Max)



OProduct structure : Silicon monolithic integrated circuit OThis product has no designed protection against radioactive rays

Typical Application Circuit

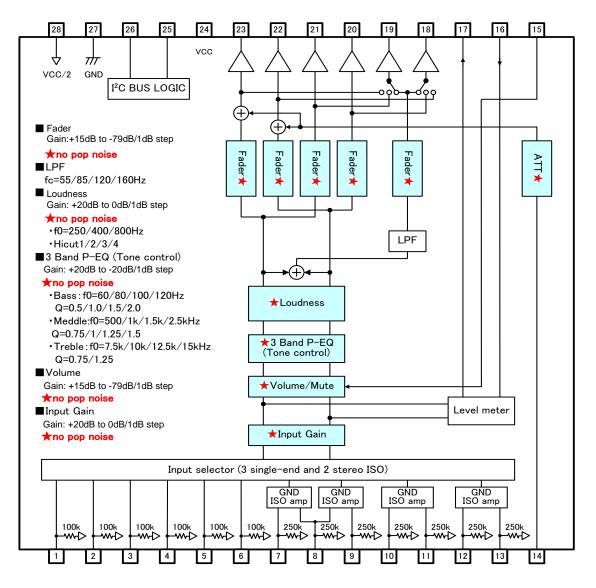


Pin Descriptions

Pin Configuration

Pin No.	Pin Name	Description	Pin No.	Pin Name	Description
1	A1	A input terminal of 1ch	15	MUTE	External compulsory mute terminal
2	A2	A input terminal of 2ch	16	LRST	Level meter reset terminal
3	B1	B input terminal of 1ch	17	LOUT	Output terminal for Level meter
4	B2	B input terminal of 2ch	18	OUTS2	Subwoofer output terminal of 2ch
5	C1	C input terminal of 1ch	19	OUTS1	Subwoofer output terminal of 1ch
6	C2	C input terminal of 2ch	20	OUTR2	Rear output terminal of 2ch
7	DP1	D positive input terminal of 1ch	21	OUTR1	Rear output terminal of 1ch
8	DN	D negative input terminal	22	OUTF2	Front output terminal of 2ch
9	DP2	D positive input terminal of 2ch	23	OUTF1	Front output terminal of 1ch
10	EP1	E positive input terminal of 1ch	24	VCC	Power supply terminal
11	EN1	E negative input terminal of 1ch	25	SCL	I ² C Communication clock terminal
12	EN2	E negative input terminal of 2ch	26	SDA	I ² C Communication data terminal
13	EP2	E positive input terminal of 2ch	27	GND	GND terminal
14	MIN	Mixing input terminal	28	FIL	VCC/2 terminal

Block Diagram



Absolute Maximum Ratings (Ta=25°C)

Parameter	Symbol	Rating	Unit
Power Supply Voltage	Vcc	10.0	V
Input Voltage	Vin	Vcc+0.3 to GND-0.3	V
Power Dissipation	Pd	1.06 (Note 1)	W
Storage Temperature	Tstg	-55 to +150	°C

(Note 1) When mounted on the standard board (70 x 70 x 1.6(mm³), derate by 8.5mW/°C for Ta above 25°C.

Thermal resistance θja = 117.6(°CW) Material : A FR4 grass epoxy board (3% or less of copper foil area) Caution: Operating the IC over the absolute maximum ratings may damage the IC. The damage can either be a short circuit between pins or an open circuit between pins and the internal circuitry. Therefore, it is important to consider circuit protection measures, such as adding a fuse, in case the IC is operated over the absolute maximum ratings.

Recommended Operating Conditions

Parameter	Symbol	Limit	Unit
Power Supply Voltage	Vcc	7.0 to 9.5	V
Temperature	Topr	-40 to +85	°C

Electrical Characteristics

(Unless otherwise noted, Ta=25°C, Vcc=8.5V, f=1kHz, VIN=1Vrms, Rg=600Ω, RL=10kΩ, A1 input, Input gain 0dB, Mute OFF, Volume 0dB, Tone control 0dB, Loudness 0dB, LPF OFF, Mixing OFF, Fader 0dB)

Voltage GainGv-1.501.5dBGv=2Channel BalanceCB-1.501.5dBCB =Total Harmonic Distortion 1 (FRONT,REAR)THD+N1-0.0010.05%Vout BW=	Conditions ignal 20log(Vout/Vin) = Gv1-Gv2
Voltage Gain Gv -1.5 0 1.5 dB Gv=2 Channel Balance CB -1.5 0 1.5 dB CB = Total Harmonic Distortion 1 (FRONT,REAR) THD+N1 - 0.001 0.05 % Vour BW=	20log(Vout/Vin) = Gv1-Gv2
Channel BalanceCB-1.501.5dBCB =Total Harmonic Distortion 1 (FRONT,REAR)THD+N1-0.0010.05%Vout BW=	= Gv1-Gv2
Total Harmonic Distortion 1 (FRONT,REAR)THD+N1-0.0010.05%Vout BW=	
(FRONT,REAR)	
	≔1Vrms =400Hz-30KHz
(SUBWOOFER) THD+N2 - 0.002 0.05 % BW=	≔1Vrms ⊧400Hz-30KHz
$\int_{-\infty}^{\infty} \frac{\text{Output Noise Voltage 1}}{(\text{FRONT,REAR})^*} \qquad \qquad$	= IHF-A
	= IHF-A
Residual Output Noise Voltage * V _{NOR} - 1.8 10 µVrms Rg =	er = -∞dB : 0Ω = IHF-A
	= 0Ω =20log(V _{OUT} /V _{IN}) = IHF-A
	Hz =100mVrms 20log(V _{CC} IN/V _{Ουτ})
Input Impedance(A, B, C) R _{IN_S} 70 100 130 kΩ	
Input Impedance (D, E) RIN_D 175 250 325 k Ω	
PMaximum Input VoltageVIM2.12.3-VrmsVIM a BW=	at THD+N(Vout)=1% =400Hz-30KHz
0 BW-	= 0Ω =20log(Vout/Vin) = IHF-A
ECommon Mode Rejection Ratio * (D, E)CMRR5065-dBXP2 CMR	and XN input and XN input R=20log(VIN/Vout) = IHF-A,[*X・・・D,E]
	t gain 0dB 100mVrms 20log(Vo∪⊤/V៲ℕ)
\Box Maximum Input Gain $G_{IN_{MAX}}$ +18 +20 +22 dB V_{IN}	t gain +20dB 100mVrms 20log(Vo∪t/Vi∖)
	N=+20dB to +1dB

Electrical Characteristics - continued

	al Characteristics - continued			Limit			
BLOCK	Parameter	Symbol	Min	Тур	Max	Unit	Conditions
MUTE	Mute Attenuation *	GMUTE	-	-105	-85	dB	Mute ON G _{MUTE} =20log(V _{OUT} /V _{IN}) BW = IHF-A
	Maximum Gain	Gv_max	13	15	17	dB	Volume = $15dB$ V _{IN} =100mVrms G _V =20log(V _{OUT} /V _{IN})
VOLUME	Maximum Attenuation *	G _{V_MIN}	-	-100	-85	dB	Volume = $-\infty dB$ G _V =20log(V _{OUT} /V _{IN}) BW = IHF-A
Š	Attenuation Set Error 1	Gv_err1	-2	0	+2	dB	GAIN & ATT=+15dB to -15dB
	Attenuation Set Error 2	G _{V_ERR2}	-3	0	+3	dB	ATT=-16dB to -47dB
	Attenuation Set Error 3	Gv_err3	-4	0	+4	dB	ATT=-48dB to -79dB
	Maximum Boost Gain	G _{B_BST}	18	20	22	dB	GAIN=+20dB f=100Hz V _{IN} =100mVrms GB=20log (V _{OUT} /V _{IN})
BASS	Maximum Cut Gain	G _{B_CUT}	-22	-20	-18	dB	GAIN=-20dB f=100Hz V _{IN} =2Vrms GB=20log (V _{OUT} /V _{IN})
	Gain Set Error	Gb_err	-2	0	+2	dB	Gain=+20dB to -20dB f=100Hz
ш	Maximum Boost Gain	G _{M_BST}	18	20	22	dB	Gain=+20dB f=1KHz V _{IN} =100mVrms GM=20log (V _{OUT} /V _{IN})
MIDDLE	Maximum Cut Gain	G м_сит	-22	-20	-18	dB	Gain=-20dB f=1KHz V _{IN} =2Vrms GM=20log (V _{OUT} /V _{IN})
	Gain Set Error	Gm_err	-2	0	+2	dB	GAIN=+20dB to -20dB f=1KHz
ш	Maximum Boost Gain	Gt_bst	18	20	22	dB	Gain=+20dB f=10kHz V _{IN} =100mVrms GT=20log (V _{OUT} /V _{IN})
TREBLE	Maximum Cut Gain	Gt_cut	-22	-20	-18	dB	Gain=-20dB f=10kHz V _{IN} =2Vrms GT=20log (V _{OUT} /V _{IN})
	Gain Set Error	Gt_err	-2	0	+2	dB	Gain=+20dB to -20dB f=10kHz
	Input Impedance	RIN_M	19	27	35	kΩ	
U	Maximum Input Voltage	Vім_м	2.0	2.2	-	Vrms	V _{IM} at THD+N(V _{OUT})=1% BW=400Hz-30KHz
MIXING	Maximum Attenuation *	G _{MX_MIN}	-	-100	-85	dB	MIX=OFF G _{MX} =20log(V _{OUT} /V _{IN}) BW=INF-A
	Maximum Gain	GMX_MAX	5	7	9	dB	ATT=+7dB G _{MX} =20log(V _{OUT} /V _{IN})
	Maximum Boost Gain	GF_BST	13	15	17	dB	Fader=15dB V_{IN} =100mVrms G_F =20log(V_{OUT}/V_{IN})
SUBWOOFER	Maximum Attenuation *	GF_MIN	-	-100	-90	dB	Fader = -∞dB GF=20log(V _{OUT} /V _{IN}) BW = IHF-A
IBWC	Gain Set Error	Gf_err	-2	0	+2	dB	Gain=+15dB to +1dB
	Attenuation Set Error 1	G_{F_ERR1}	-2	0	+2	dB	ATT=-1dB to -15dB
FADER /	Attenuation Set Error 2	GF_ERR2	-3	0	+3	dB	ATT=-16dB to -47dB
=AC	Attenuation Set Error 3	G _{F_ERR3}	-4	0	+4	dB	ATT=-48dB to -79dB
	Output Impedance	Rout	-	-	50	Ω	V _{IN} =100mVrms
	Maximum Output Voltage	V _{OM}	2	2.2	-	Vrms	THD+N=1% BW=400Hz-30KHz

Electrical Characteristics - continued

X				Limit			
BLOCK	Parameter	Symbol	Min	Тур	Max	Unit	Conditions
LOUDNESS	Maximum Gain	Gl_max	17	20	23	dB	Gain 20dB Viℕ=100mVrms GL=20log(V _{OUT} /Viℕ)
LOL	Gain Set Error	Gl_err	-2	0	+2	dB	GAIN=+20dB to +1dB
Level meter	Maximum Output Voltage	VL_MAX	2.8	3.1	3.5	V	
Level	Output Offset Voltage	Vl_off	-	0	100	mV	

VP-6690A (Average value detection, effective value display) filter by Matsushita Communication is used for * measurement. Phase between input / output is same.

Typical Performance Curves

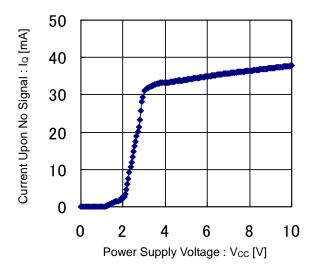


Figure 1. Circuit Current (No Signal) vs Power Supply Voltage

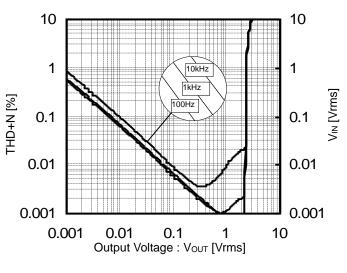


Figure 2. THD+N vs Output Voltage

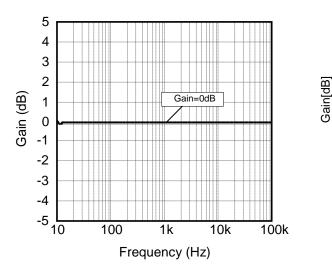


Figure 3. Gain vs Frequency

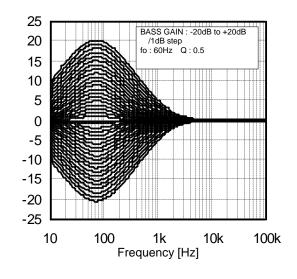
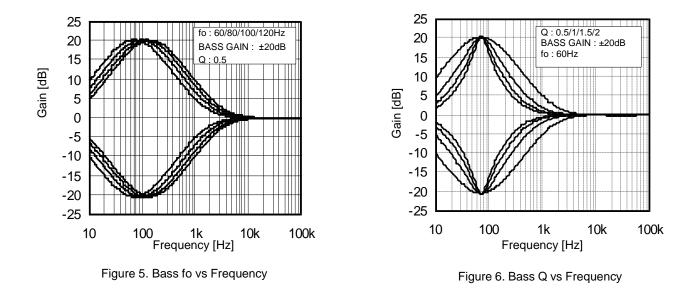
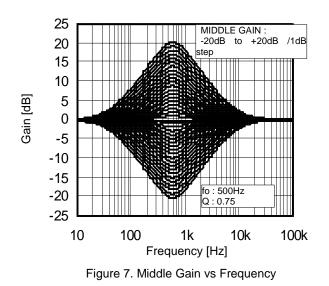


Figure 4. Bass Gain vs Frequency





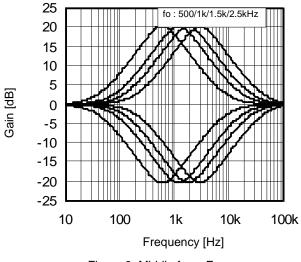
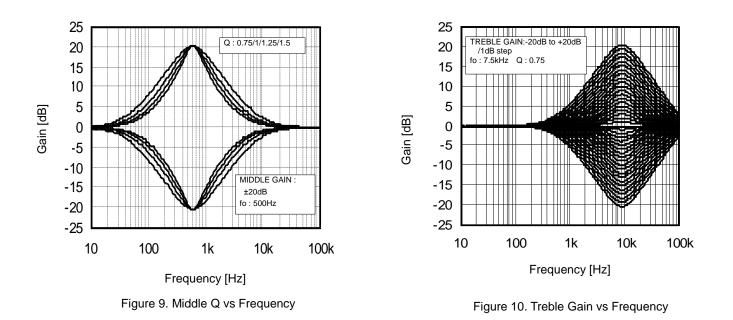


Figure 8. Middle fo vs Frequency



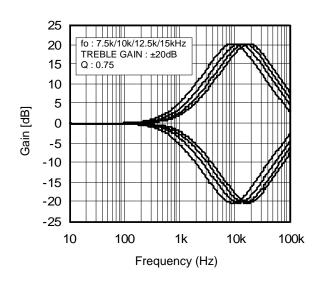


Figure 11. Treble fo vs Frequency

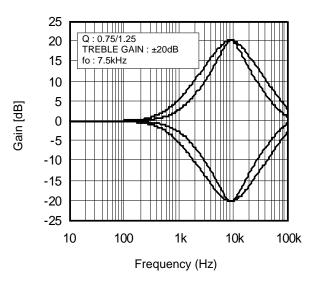


Figure 12. Treble Q vs Frequency

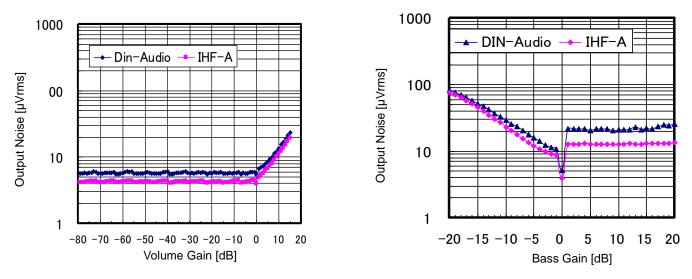
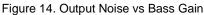


Figure 13. Output Noise vs Volume Gain



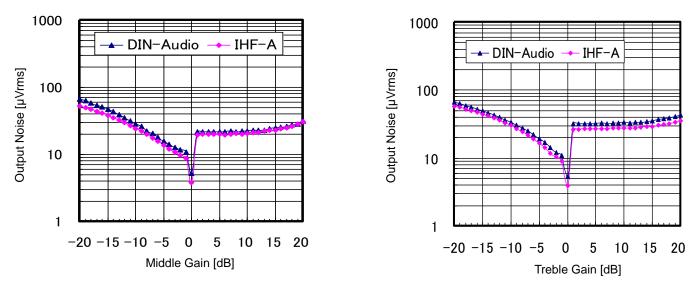


Figure 15. Output Noise vs Middle Gain

Figure 16. Output Noise vs Treble Gain

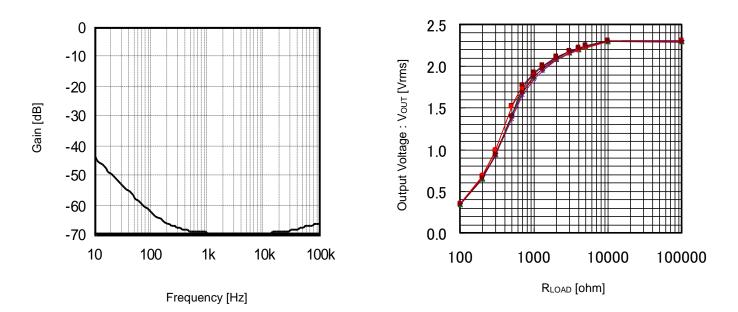


Figure 17. CMRR vs Frequency

Figure 18. Output Voltage vs R_{LOAD}

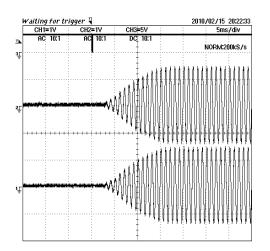


Figure 19. Advanced Switch 1

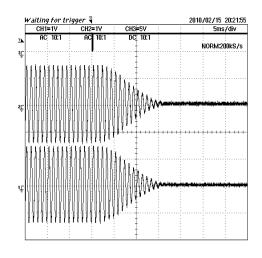


Figure 20. Advanced Switch 2

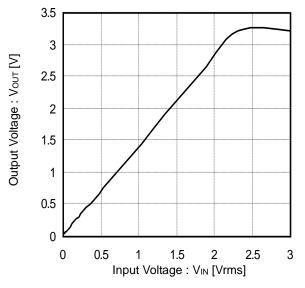


Figure 21. Output Voltage vs Level Meter V_{IN}

Timing Chart

CONTROL SIGNAL SPECIFICATION

(1) Electrical Specifications and Timing for Bus Lines and I/O Stages

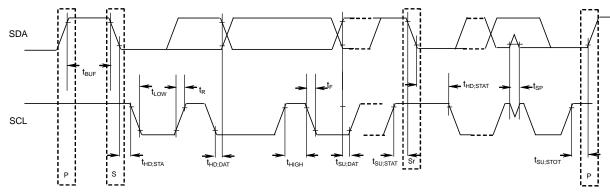


Figure 22. Definition of Timing on the I²C-bus

Table 1 Characteristics of the SDA and SCL bus lines for I^2 C-bus devices (Unless otherwise noted, Ta=25°C, V_{CC}=8.5V)

	Parameter	Symbol	Fast-mode	Unit		
	Falameter	Symbol	Min	Max	Onit	
1	SCL clock frequency	f _{SCL}	0	400	kHz	
2	Bus free time between a STOP and START condition	t _{BUF}	1.3	-	μS	
3	Hold time (repeated) START condition. After this period, the first clock	tuport	0.6			
3	pulse is generated	thd;sta	0.6	-	μS	
4	LOW period of the SCL clock	t _{LOW}	1.3	-	μS	
5	HIGH period of the SCL clock	tнigн	0.6	-	μS	
6	Set-up time for a repeated START condition	tsu;sta	0.6	-	μS	
7	Data hold time	thd;dat	0.06 ^(Note)	-	μS	
8	Data set-up time	t _{SU;DAT}	120	-	ns	
9	Set-up time for STOP condition	tsu;sто	0.6	-	μS	

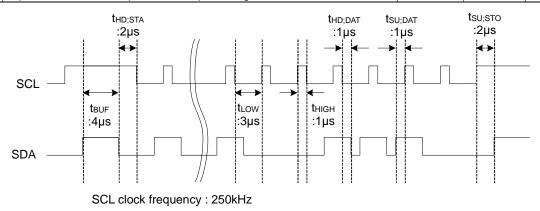
All values referred to VIH Min and VIL Max Levels (see Table 2).

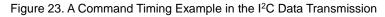
(Note) The device must internally provide a hold time of at least 300 ns for the SDA signal (referred to the VIH Min of the SCL signal) in order to bridge the undefined region of the falling edge of SCL.

About 7(t_{HD;DAT}), 8(t_{SU;DAT}), make the setup in which the margin is fully in.

Table 2 Characteristics of the SDA and SCL I/O stages for I²C-bus devices

	Parameter	Symbol	Fast-mode	Unit	
	Faldinelei	Symbol	Min	Max	Unit
10	LOW level input voltage	VIL	-0.3	+1	V
11	HIGH level input voltage	VIH	2.3	5	V
12	Pulse width of spikes which must be suppressed by the input filter.	t _{SP}	0	50	ns
13	LOW level output voltage: at 3mA sink current	Vol1	0	0.4	V
14	Input current each I/O pin with an input voltage between 0.4V and 4.5V.	li li	-10	+10	μA





(2) <u>I²C BUS FORMAT</u>

	MSB	LSB		MSB	LSB		MSB		LSB		
S	Slave Add	ress	Α	Select Add	dress	Α		Data		Α	Р
1bit	8bit	8bit 11		8bi	t	1bit		8bit		1bit	1bit
	S = Start conditions (Recognition of start bit)										
	Slave Ac	dress	= Recognition of slave address. 7 bits in upper order are voluntary.								
			The least significant bit is "L" due to writing.								
	А		= ACKNOWLEDGE bit (Recognition of acknowledgement)								
Select Address = Select every of volume, bass and treble.											
	Data = Data on every volume and tone.										
	P = Stop condition (Recognition of stop bit)										

(3) <u>I²C BUS Interface Protocol</u>

(a)) E	Basic Form								
	S	Slave A	ddress	Α	Select	Address	Α	Data	Α	Ρ
		MSB	LSB		MSB	LSB	Ν	1SB	LSB	

(b) Automatic increment (Select Address increases (+1) according to the number of data.

	S	Slave Address	А	Select Address	А	Data1	А	Data2	А		DataN	А	Ρ
-		-	-	ASB LSE		SB LSB		-	SB	MS	SB LS	В	

(Example) ①Data1 shall be set as data of address specified by Select Address.
 ②Data2 shall be set as data of address specified by Select Address +1.

3DataN shall be set as data of address specified by Select Address +N-1.

- (c) Configuration unavailable for transmission (In this case, only Select Address1 is set.
- S Slave Address A Select Address1 A Data A Select Address 2 A Data A P

MSB	LSB	MSB	LSB	MSB	LSB	MSB	LSB MSB	LSB
(Note)	If any da	ata is transı	mitted as S	elect Ac	ldress 2	next to	data, it is recognized	
	as dat	a, not as S	elect Addre	ess 2.			-	

(4) Slave Address

ſ	MSB							LSB	
	A6	A5	A4	A3	A2	A1	A0	R/W	
	1	0	0	0	0	0	0	0	80H

(5) <u>Select Address & Data</u>

Items	Select Address	MSB			Da	ata			LSB		
items	(hex)	D7	D6	D5	D4	D3	D2	D1	D0		
Initial setup 1	01	Advanced switch ON/OFF	0	time o Gain/∖	ed switch f Input /olume er/Loudnes xing	0	1	Advanced of I	switch time ⁄lute		
Initial setup 2	02	LPF Phase	Level Meter RESET		er Output lect	0	Su	ibwoofer LP	F fc		
Initial setup 3	03	0	0	0	Loudn	less fo	0	0	1		
Input Selector	05	Full-diff Type	0	0			Input selector				
Input gain	06	Mute ON/OFF	0	0	Input Gain						
Volume gain	20			١	Volume Gain / Attenuation						
Fader 1ch Front	28				Fader Gain / Attenuation						
Fader 2ch Front	29	Fader Gain / Attenuation									
Fader 1ch Rear	2A				Fader Gain / Attenuation						
Fader 2ch Rear	2B				Fader Gain	/ Attenuatio	n				
Fader Subwoofer	2C				Fader Gain	/ Attenuatio	n				
Mixing	30				Mixing Gain	/ Attenuatio	n				
Bass setup	41	0	0	Bas	s fo	0	0	Ba	ss Q		
Middle setup	44	0	0	Mido	lle fo	0	0	Mide	dle Q		
Treble setup	47	0	0	Treb	le fo	0	0	0	Treble Q		
Bass gain	51	Bass Boost/ Cut	0	0			Bass Gain				
Middle gain	54	Middle Boost/ Cut	0	0	Middle Gain						
Treble gain	57	Treble Boost/ Cut	0	0	Treble Gain						
Loudness Gain	75	0	Loudne	ss Hicut		L	oudness Ga	lin			
System Reset	FE	1	0	0	0	0	0	0	1		

Advanced switch

Note

- 1. The Advanced Switch works in the latch part while changing from one function to another.
- 2. Upon continuous data transfer, the Select Address rolls over because of the automatic increment function, as shown below.

$$\rightarrow 01 \rightarrow 02 \rightarrow 03 \rightarrow 05 \rightarrow 06 \rightarrow 20 \rightarrow 28 \rightarrow 29 \rightarrow 2A \rightarrow 2B \rightarrow 2C \rightarrow 30 \rightarrow 41 \rightarrow 44 \rightarrow 47 \rightarrow 51 \rightarrow 54 \rightarrow 57 \rightarrow 75 \rightarrow 75$$

- 3. Advanced switch is not used for functions of input selector and subwoofer output select etc. Therefore, please turn on MUTE when changing the settings of this side of a set.
- 4. When using Mute function of this IC at the time of changing input selector, please switch mute ON/OFF while waiting for advanced-mute time.

Time	MSB	ISB Advanced switch time of Mute LSE									
TIME	D7	D6	D5	D4	D3	D2	D1	D0			
0.6msec	Advanced	Advanced		switch time	0	1	0	0			
1.0msec	Switch	0	of Input gain/Volume				0	1			
1.4msec	ON/OFF		Tone/Fade	r/Loudness	0	1	1	0			
3.2msec			/Miz	xing			1	1			

	Time	MSB	gain/Volume/Tone/Fader/Loudness/Mixing							
		D7	D6	D5	D4	D3	D2	D1	D0	
	4.7 msec	Advanaad	0	0	0	0				
	7.1 msec	Advanced Switch		0	1		1	Advanced switch Time of Mute		
ĺ	11.2 msec	ON/OFF		1	0		1			
Ī	14.4 msec	UN/OFF		1	1					

Mode	MSB		LSB					
Mode	D7	D6	D5	D4	D3	D2	D1	D0
OFF	0	0	Advanced switch time of Input gain/Volume Tone/Fader/Loudness /Mixing		0	1	Advanced switch	
ON	1	0			0	-	Time o	of Mute

Select address 02(hex)								
fc	MSB		Su	bwoof	er LPF	⁻ fc		LSB
IC IC	D7	D6	D5	D4	D3	D2	D1	D0
OFF						0	0	0
55Hz		Level	Subwoofer Output			0	0	1
85Hz	LPF				0	0	1	0
120Hz	Phase	Meter RESET	Se	lect	0	0	1	1
160Hz		RESET				1	0	0
Prohibition							Other setting)

Mode	MSB Subwoofer Output Select								
Mode	D7	D6	D5	D4	D3	D2	D1	D0	
LPF	LPF Phase	Level Meter RESET	0	0		i			
Front			0	1	0	Subwoofer LPF fc			
Rear			1	0					
Prohibition		INCOL I	1	1					

Mode	MSB	B Level Meter RESET								
Mode	D7	D6	D5	D4	D3	D2	D1	D0		
HOLD	LPF	0	Subwoofer output select		0					
RESET	Phase	1			0	Subwoofer LPF fc		- IC		

Phase	MSB				LSB				
Thase	D7 D6			D4	D3	D2	D1	D0	
0°	0	Level	Subwoof	fer output	0	C	Subwoofer LDE fo		
180°	1	Meter RESET	se	lect	0	Subwoofer LPF fc		- 10	

: Initial Condition

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Select address 03(hex)

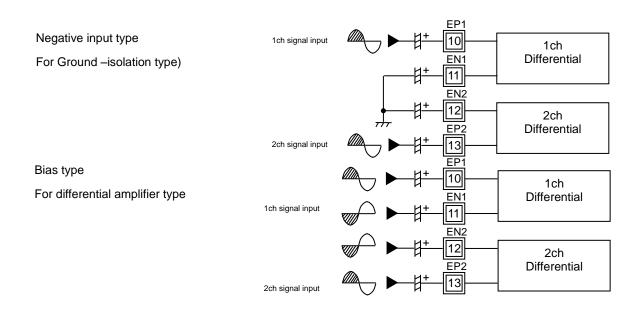
f0	MSB			Loudr	less fo			LSB
10	D7	D6	D5	D4	D3	D2	D1	D0
250Hz		_	0	0	0		0	
400Hz	0			0	1	0		4
800Hz	0	0		1	0			1
Prohibition				1	1			

Select address 05(hex)

Mada			MSB			Input S	elector			LSB	
Mode	OUTF1	OUTF2	D7	D6	D5	D4	D3	D2	D1	D0	
A	A1	A2				0	0	0	0	0	
В	B1	B2				0	0	0	0	1	
С	C1	C2				0	0	0	1	0	
D single	DP1	DP2				0	0	0	1	1	
E1 single	EP1	EN1	Full-diff		0	1	0	1	0		
E2 single	EN2	EP2		0	0	0	1	0	1	1	
A diff	A1	B1	bias type select		0	0	1	1	1	1	
C diff	B2	C2	Select			1	0	0	0	0	
D diff	DP1	DP2				0	0	1	1	0	
E full diff	EP1	EP2				0	1	0	0	0	
Inp	ut SHORT					0	1	0	0	1	
Prohibition					-	Other setting					

Input SHORT : The input impedance of each input terminal is lowered from $100k\Omega(TYP)$ to $6 k\Omega(TYP)$. (For quick charge of coupling capacitor)

Mode	MSB			Full-diff Bia	LSB			
	D7	D6	D5	D4	D3	D2	D1	D0
Negative Input	0	0	0		1	nnut Calaata	-	
Bias	0	0	Input Selector					



Gain	MSB			Input	Gain			LSE
Galli	D7	D6	D5	D4	D3	D2	D1	D0
0dB				0	0	0	0	0
1dB				0	0	0	0	1
2dB				0	0	0	1	0
3dB				0	0	0	1	1
4dB				$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	0	0		
5dB					0	1		
6dB					1	0		
7dB					1	1		
8dB					0	0		
9dB				0	1	0	0	1
10dB				0	1	0	1	0
11dB	Mute	0	0	0	1	0	1	1
12dB	ON/OFF	0	0	0	1	1	0	0
13dB				0	1	1	0	1
14dB				0	1	1	1	0
15dB				0	1	1	1	1
16dB				1	0	0	0	0
17dB				1	0	0	0	1
18dB				1	0	0	1	0
19dB				1	0	0	1	1
20dB				1	0	1	0	0
				1	1	0	1	1
Prohibition				:	:		:	:
				1	1	1	1	1

Mode	MSB		Γ	/lute C	N/OF	F		LSB
Mede	D7	D6	D5	D4	D3	D2	D1	D0
OFF	0	0	0			Input Gain		
ON	1	0	0			Input Gain		

BD37534FV

	MSB	Vo	ol, Fad	ler Gai	n / Atte	enuati	on	LSB
Gain & ATT	D7	D6	D5	D4	D3	D2	D1	D0
	0	0	0	0	0	0	0	0
	0	0	0	0	0	0	0	1
Prohibition	:	:	:	:	:	:	:	:
	0	1	1	1	0	0	0	0
15dB	0	1	1	1	0	0	0	1
14dB	0	1	1	1	0	0	1	0
13dB	0	1	1	1	0	0	1	1
:	:	:	:	:	:	:	:	:
-77dB	1	1	0	0	1	1	0	1
-78dB	1	1	0	0	1	1	1	0
-79dB	1	1	0	0	1	1	1	1
	1	1	0	1	0	0	0	0
Prohibition	:	:	:	:	:	•	:	:
	1	1	1	1	1	1	1	0
-∞dB	1	1	1	1	1	1	1	1

Select address 20, 28, 29, 2A, 2B, 2C (hex)

Select address 30(hex)

Gain & ATT	MSB		Mixing) Gain	/ Atten	uation		LSB
Gaill & ATT	D7	D6	D5	D4	D3	D2	D1	D0
	0	0	0	0	0	0	0	0
	0	0	0	0	0	0	0	1
Prohibition	:	:	:	:	:	:	:	:
	0	1	1	1	1	0	0	0
7dB	0	1	1	1	1	0	0	1
6dB	0	1	1	1	1	0	1	0
5dB	0	1	1	1	1	0	1	1
:	:	:	:	:	:	:	:	:
-77dB	1	1	0	0	1	1	0	1
-78dB	1	1	0	0	1	1	1	0
-79dB	1	1	0	0	1	1	1	1
	1	1	0	1	0	0	0	0
Prohibition	:	:	:	:	:	:	:	:
	1	1	1	1	1	1	1	0
MIX OFF	1	1	1	1	1	1	1	1

Select address 41(hex)

Q factor	MSB		Ba	ass C	LSB			
QTACIO	D7	D6	D5	D4	D3	D2	D1	D0
0.5							0	0
1.0	0	0	Poo	a fa	0	0	0	1
1.5	0	0	Das	ss fo	0	0	1	0
2.0							1	1

fo	MSB			Bass	fo LS			LSB
fo	D7	D6	D5	D4	D3	D2	D1	D0
60Hz			0	0				
80Hz		0	0	1	0	0	Ba	ass actor
100Hz	0	0	1	0	0	0	Q fa	actor
120Hz	7		1	1				

Select address 44(hex)

Q factor	MSB		Mic	ddle	Q facto	or		LSB
	D7	D6	D5	D4	D3	D2	D1	D0
0.75							0	0
1.0	0	0	Mida	dle fo	0	0	0	1
1.25	0	0	WILCO		0	0	1	0
1.5							1	1

fo	fo MSB			Middle fo LSI					
10	D7	D6	D5	D4	D3	D2	D1	D0	
500Hz			0	0					
1kHz	_	0	0	1		0	Mic	ldle actor	
1.5kHz	0	0	1	0	0	0	Q fa	actor	
2.5kHz			1	1					

Select address 47 (hex)

Q factor	MSB		Tr∉	eble	Q facto	or	LSB	
QIACIOI	D7	D6	D5	D4	D3	D2	D1	D0
0.75	0	0	Trok	la fa	0	0	0	0
1.25	0	0	Treble fo		0	0	0	1

fo	MSB		Treble fo LSB						
fo	D7	D6	D5	D4	D3	D2	D1	D0	
7.5kHz			0	0					
10kHz	0	0	0	1	0	0	0	Treble	
12.5kHz	0	0	1	0	0	0	0	Q factor	
15kHz			1	1					

: Initial condition

Cain	MSB	E	3ass/№	1iddle/	Treble	Gain		LSE
Gain	D7	D6	D5	D4	D3	D2	D1	D
0dB				0	0	0	0	C
1dB				0	0	0	0	1
2dB				0	0	0	1	C
3dB				0	0	0	1	1
4dB				0	0	1	0	0
5dB				0	0	1	0	1
6dB				0	0	1	1	C
7dB				0	0	1	1	1
8dB				0	1	0	0	C
9dB				0	1	0	0	1
10dB	Bass/			0	1	0	1	C
11dB	Middle/			0	1	0	1	1
12dB	Treble	0	0	0	1	1	0	C
13dB	Boost	•		0	1	1	0	1
14dB	/cut			0	1	1	1	C
15dB				0	1	1	1	1
16dB				1	0	0	0	C
17dB				1	0	0	0	1
18dB				1	0	0	1	C
19dB				1	0	0	1	1
20dB				1	0	1	0	C
				1	0	1	0	1
Prohibition				:	:	:	:	:
1 Tornollon				1	1	1	1	(
				1	1	1	1	1

Mode	MSB	Bas	ss/Mid	dle/Tre	eble Bo	oost/C	ut	LSB
Mode	D7 D6 D5 D4 D3 D2 D1							D0
Boost	0	0	0		Base/	Middle/Troble	Cain	
Cut	1	0	0 0 Bass/Middle/Treble Gain					

Γ

Select address 75 (hex)									
Mode	MSB Loudness Hicut									
wode	D7	D6	D5	D4	D3	D2	D1	D0		
Hicut1		0	0							
Hicut2		0	1				in			
Hicut3	0	1	0	Loudness Gain						
Hicut4		1	1							

Gain	MSB		L	oudne	ess Ga	in		LSB
Gain	D7	D6	D5	D4	D3	D2	D1	D0
0dB			•	0	0	0	0	0
1dB				0	0	0	0	1
2dB				0	0	0	1	0
3dB				0	0	0	1	1
4dB				0	0	1	0	0
5dB				0	0	1	0	1
6dB				0	0	1	1	0
7dB				0	0	1	1	1
8dB				0	1	0	0	0
9dB				0	1	0	0	1
10dB				0	1	0	1	0
11dB				0	1	0	1	1
12dB	0	Loudne	ss Hicut	0	1	1	0	0
13dB				0	1	1	0	1
14dB				0	1	1	1	0
15dB				0	1	1	1	1
16dB				1	0	0	0	0
17dB				1	0	0	0	1
18dB				1	0	0	1	0
19dB				1	0	0	1	1
20dB				1	0	1	0	0
				1	0	1	0	1
Prohibition				:	:	:	:	:
				1	1	1	1	1

: Initial condition

Г

(6) About Power ON Reset

Built-in IC initialization is made during power ON of the supply voltage. Please send initial data to all addresses at supply voltage on. Also, please turn ON MUTE at the set side until initial data is sent.

Deremeter	Symbol	Symbol		Linit	Conditions		
Parameter	Symbol	Min	Тур	Max	Unit	Conditions	
Rise Time of VCC	t _{RISE}	33	-	-	µsec	V_{CC} rise time from 0V to 5V	
VCC Voltage of Release Power ON Reset	V _{POR}	-	4.1	-	V		

(7) About External Compulsory Mute Terminal

It is possible to forcibly set MUTE externally by setting the input voltage at the MUTE terminal.

Mute Voltage Condition	Mode
GND to 1.0V	MUTE ON
2.3V to Vcc	MUTE OFF

Establish the mode of MUTE using a defined voltage condition.

Application Information

1. Function and Specifications

Function			Specifications								
	(Stereo input)										
	Single-End/Diff/Full-Diff										
	(Possible to set the number of single-end/diff/full-diff as follows:)										
		Single-End	Differential	Full-Differential							
Input selector	Mode 1	0	3	1							
	Mode 2 Mode 3	<u>1</u> 3	2	1							
	Mode 4	4	0	1							
	Mode 5 Mode 6	<u>5</u> 6	1	0							
	NOUE 0	-	ombination of inp	•	l						
Input	• +20dB to	0dB (1dB step))								
gain	 Possible to use "Advanced switch" for prevention of switching noise 										
Mute	 Possible t 	o use "Advanc	ed switch" for pre	evention of switching	noise.						
Volume	・+15dB to	-79dB (1dB ste	ep), -∞dB								
	 Possible t 	o use "Advanc	ed switch" for pre	evention of switching	noise.						
	・+20dB to	-20dB (1dB ste	ep)								
Bass	· Q=0.5, 1, 1.5, 2										
Dass	• fo=60, 80, 100, 120Hz										
	Possible to use "Advanced switch" when changing gain										
	+20dB to -20dB (1dB step)										
Middle	· Q=0.75, 1, 1.25, 1.5										
Widdle	• fo=500, 1k, 1.5k 2.5kHz										
	 Possible t 	o use "Advanc	ed switch" when	changing gain							
	 +20dB to -20dB (1dB step) 										
Treble	・Q=0.75, 1	.25									
Troble	• fo=7.5k, 10k, 12.5k, 15kHz										
	Possible to use "Advanced switch" when changing gain										
Fader	• +15dB to	-79dB(1dB ste	p), -∞dB								
	 Possible t 	o use "Advanc	ed switch" for pre	evention of switching	noise.						
	20dB to 0dB(1dB step)										
Loudness	· fo=250/40	0/800Hz									
	 Possible t 	o use "Advanc	ed switch" for pre	evention of switching	noise.						
LPF	• fc=55/85/	120/160Hz, pas	SS								
	Phase shi	ft (0°/180°)									
	・ Monaural	input									
Mixing	• +7dB to -	79dB (1dB step	o), -∞dB								
	Possible t	o use "Advanc	ed switch" for pre	evention of switching	noise.						
Level	• I ² C BUS o	control									
meter	・DC Outpu	t									

2. Volume / Fader Volume / Mixing Attenuation of the Details

2.	Volume	: / rau		Jume		ing A	ttenu	ation	or the	e Deta	lis								
((dB)	D7	D6	D5	D4	D3	D2	D1	D0		(dB)	D7	D6	D5	D4	D3	D2	D1	D0
	+15	0	1	1	1	0	0	0	1		-33	1	0	1	0	0	0	0	1
	+14	0	1	1	1	0	0	1	0		-34	1	0	1	0	0	0	1	0
	+13	0	1	1	1	0	0	1	1		-35	1	0	1	0	0	0	1	1
	+12	0	1	1	1	0	1	0	0		-36	1	0	1	0	0	1	0	0
_	+11	0	1	1	1	0	1	0	1		-37	1	0	1	0	0	1	0	1
	+10	0	1	1	1	0	1	1	0		-38	1	0	1	0	0	1	1	0
	+9	0	1	1	1	0	1	1	1		-39	1	0	1	0	0	1	1	1
-	+8	0	1	1	1	1	0	0	0		-40	1	0	1	0	1	0	0	0
-	+7	0	1	1	1	1	0	0	1		-41	1	0	1	0	1	0	0	1
-	+6	0	1	1	1	1	0	1	0		-42	1	0	1	0	1	0	1	0
-	+5	0	1	1	1	1	0	1	1		-43	1	0	1	0	1	0	1	1
-	+4	0	1	1	1	1	1	0	0		-44	1	0	1	0	1	1	0	0
-	+3	0	1	1	1	1	1	0	1		-45	1	0	1	0	1	1	0	1
-	+2	0	1	1	1	1	1	1	0		-46	1	0	1	0	1	1	1	0
<u> </u>	+1	0	1	1	1	1	1	1	1		-40	1	0	1	0	1	1	1	1
-	0	1	0	0	0	0	0	0	0		-48	1	0	1	1	0	0	0	0
-	-1	1	0	0	0	0	0	0	1		-49	1	0	1	1	0	0	0	1
	-2	1	0	0	0	0	0	1	0		-50	1	0	1	1	0	0	1	0
	-3	1	0	0	0	0	0	1	1		-51	1	0	1	1	0	0	. 1	1
	-4	1	0	0	0	0	1	0	0		-52	1	0	1	1	0	1	0	0
	-5	1	0	0	0	0	1	0	1		-53	1	0	1	1	0	1	0	1
	-6	1	0	0	0	0	1	1	0		-54	1	0	1	1	0	1	1	0
	-7	1	0	0	0	0	1	1	1		-55	1	0	1	1	0	1	1	1
	-8	1	0	0	0	1	0	0	0		-56	1	0	1	1	1	0	0	0
	-9	1	0	0	0	1	0	0	1		-57	1	0	1	1	1	0	0	1
	-10	1	0	0	0	1	0	1	0		-58	1	0	1	1	1	0	1	0
	-11	1	0	0	0	1	0	1	1		-59	1	0	1	1	1	0	1	1
	-12	1	0	0	0	1	1	0	0		-60	1	0	1	1	1	1	0	0
	-13	1	0	0	0	1	1	0	1		-61	1	0	1	1	1	1	0	1
	-14	1	0	0	0	1	1	1	0		-62	1	0	1	1	1	1	1	0
	-15	1	0	0	0	1	1	1	1		-63	1	0	1	1	1	1	1	1
	-16	1	0	0	1	0	0	0	0		-64	1	1	0	0	0	0	0	0
	-17	1	0	0	1	0	0	0	1		-65	1	1	0	0	0	0	0	1
	-18	1	0	0	1	0	0	1	0		-66	1	1	0	0	0	0	1	0
	-19	1	0	0	1	0	0	1	1		-67	1	1	0	0	0	0	1	1
	-20	1	0	0	1	0	1	0	0		-68	1	1	0	0	0	1	0	0
	-21	1	0	0	1	0	1	0	1		-69	1	1	0	0	0	1	0	1
	-22	1	0	0	1	0	1	1	0		-70	1	1	0	0	0	1	1	0
	-23	1	0	0	1	0	1	1	1		-71	1	1	0	0	0	1	1	1
	-24	1	0	0	1	1	0	0	0		-72	1	1	0	0	1	0	0	0
	-25	1	0	0	1	1	0	0	1		-73	1	1	0	0	1	0	0	1
	-26	1	0	0	1	1	0	1	0		-74	1	1	0	0	1	0	1	0
	-27	1	0	0	1	1	0	1	1		-75	1	1	0	0	1	0	1	1
	-28	1	0	0	1	1	1	0	0		-76	1	1	0	0	1	1	0	0
	-29	1	0	0	1	1	1	0	1		-77	1	1	0	0	1	1	0	1
	-30	1	0	0	1	1	1	1	0		-78	1	1	0	0	1	1	1	0
	-31	1	0	0	1	1	1	1	1		-79	1	1	0	0	1	1	1	1
	-32	1	0	1	0	0	0	0	0		-∞	1	1	1	1	1	1	1	1
						-	-		-										

Mixing Adjustable range is +7dB to -∞dB.

: Initial condition

- (1) About Level Meter
 - (a) The Operation of Circuit

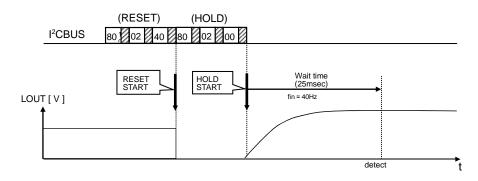
Level meter is a function which outputs a DC voltage proportional to the magnitude of the sound signal. It detects the peak level of the signal and retains that peak level, so that it is possible to monitor the magnitude of the signal by resetting the DC voltage within a suitable interval.

- (b) The Way to Reset Level Meter Output There are two kinds of methods of the reset.
 - ① Please send reset data through the I²C BUS.To reset output of level meter: Send D6 = "1" of select address 02(hex).

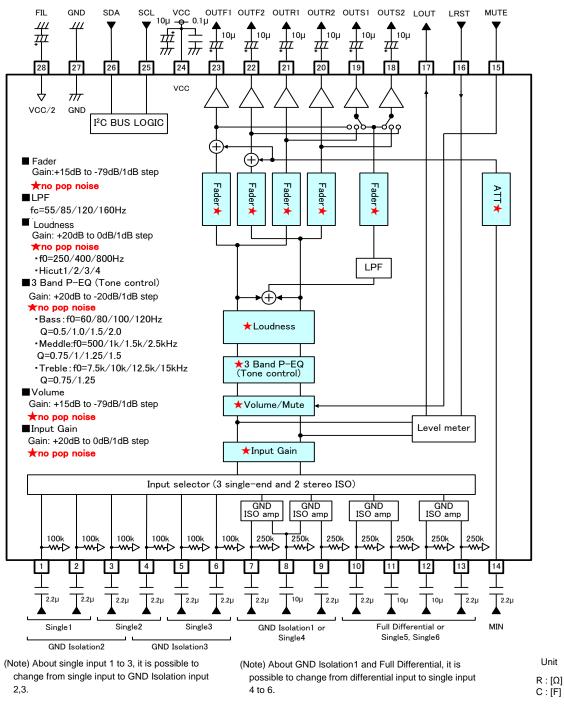
To cancel output reset of level meter (HOLD): Send D6 = "0" of select address 02(hex).

- Please control external terminal(LRST)
 To reset output of level meter: LRST terminal set to HIGL level.
 To cancel output reset of level meter (HOLD): LRST terminal set to LOW level.
- (c) The Settings About Period of Reset Peak hold operation will start after HOLD data is transmitted. Set the WAIT time after HOLD data transmission according to the frequency bandwidth detected.
 WAIT time must be set to a minimum of one cycle over the detected frequency bandwidth.
 Ex. Detected frequency bandwidth is above 40Hz, "40Hz = 25ms = WAIT time."

Transmission Example by I²C BUS



3. Application Circuit



Notes on wiring

①Please connect the decoupling capacitor of the power supply in the shortest possible distance to GND.

②GND lines should be one-point connected.

③Wiring pattern of the digital signals should be as far away as possible from those of the analog signals to prevent crosstalk.
 ④If possible, lines of SCL and SDA of I²C BUS should not be parallel.

- The lines should be shielded, if they are adjacent to each other.
- ⑤If possible, analog input lines should not be parallel.

The lines should be shielded, if they are adjacent to each other.

Power Dissipation

About the thermal design of the IC

Characteristics of the IC are affected by the temperature at which it is used. Exceeding absolute maximum ratings may degrade and destroy the device. Careful consideration must be given to the heat of the IC from the two standpoints of immediate damage and long-term reliability of operation.

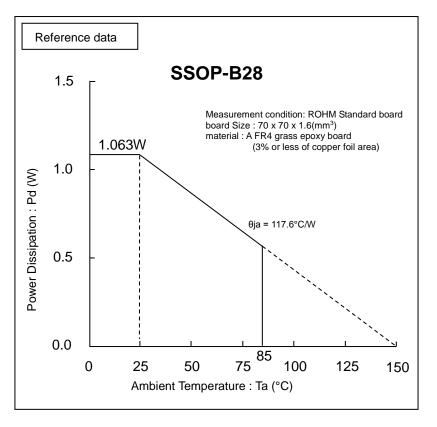


Figure 24. Temperature Derating Curve

(Note) Values are actual measurements and are not guaranteed.

Power dissipation values vary according to the board on which the IC is mounted.

I/O Equivalent Circuits

Equivale	Equivalent Circuits												
Terminal No.	Terminal Name	Terminal Voltage	Equivalent Circuit	Terminal Description									
1 2 3 4 5 6	A1 A2 B1 B2 C1 C2	4.25	VCC VOC VOC VO VO VO VO VO VO VO VO VO VO	A terminal for signal input. The input impedance is 100kΩ (Typ).									
7 8 9 10 11 12 13	DP1 DN DP2 EP1 EN1 EN2 EP2	4.25		Input terminal available to Single/Differential mode. The input impedance is 250kΩ (Typ).									
15	MUTE	-	VCC VCC VCC VCC VCC VCC VCC VCC	A terminal for external compulsory mute. If terminal voltage is High level, the mute is OFF. And if the terminal voltage is Low level, the mute is ON.									
18 19 20 21 22 23	OUTS2 OUTS1 OUTR2 OUTR1 OUTF2 OUTF1	4.25		A terminal for Fader and Subwoofer output.									
17	LOUT	0 to 3.3	VCC GND GND	A terminal for level meter output Output impedance is 10kΩ (Typ).									

Values in the pin explanation and input/output equivalent circuit are for reference purposes only. These are not guaranteed values.

I/O Equivalent Circuits – continued

Terminal No.	Terminal Name	Terminal Voltage	Equivalent Circuit	Terminal Description
				Power supply terminal.
24	VCC	8.5		
25	SCL	-	VCC GND GND	A terminal for clock input of I ² C BUS communication.
26	SDA	-	VCC The second s	A terminal for data input of I ² C BUS communication.
27	GND	0		Ground terminal.
28	FIL	4.25		Voltage for reference bias of analog signal system. The simple precharge circuit and simple discharge circuit for an external capacitor are built in.
14	MIN	4.25	VCC VCC VCC VCC VCC VCC VCC VCC VCC VCC	A terminal for signal input The input impedance is 27kΩ(Typ).
16	LRST	0	VCC VCC VCC VCC VCC VCC VCC VCC	A terminal for signal input The input impedance is 250kΩ(Typ).

Values in the pin explanation and input/output equivalent circuit are for reference purposes only. These are not guaranteed values.

Operational Notes

1. Reverse Connection of Power Supply

Connecting the power supply in reverse polarity can damage the IC. Take precautions against reverse polarity when connecting the power supply, such as mounting an external diode between the power supply and the IC's power supply pins.

2. Power Supply Lines

Design the PCB layout pattern to provide low impedance supply lines. Separate the ground and supply lines of the digital and analog blocks to prevent noise in the ground and supply lines of the digital block from affecting the analog block. Furthermore, connect a capacitor to ground at all power supply pins. Consider the effect of temperature and aging on the capacitance value when using electrolytic capacitors.

3. Ground Voltage

Ensure that no pins are at a voltage below that of the ground pin at any time, even during transient condition.

4. Ground Wiring Pattern

When using both small-signal and large-current ground traces, the two ground traces should be routed separately but connected to a single ground at the reference point of the application board to avoid fluctuations in the small-signal ground caused by large currents. Also ensure that the ground traces of external components do not cause variations on the ground voltage. The ground lines must be as short and thick as possible to reduce line impedance.

5. Thermal Consideration

Should by any chance the power dissipation rating be exceeded the rise in temperature of the chip may result in deterioration of the properties of the chip. In case of exceeding this absolute maximum rating, increase the board size and copper area to prevent exceeding the Pd rating.

6. Recommended Operating Conditions

These conditions represent a range within which the expected characteristics of the IC can be approximately obtained. The electrical characteristics are guaranteed under the conditions of each parameter.

7. Inrush Current

When power is first supplied to the IC, it is possible that the internal logic may be unstable and inrush current may flow instantaneously due to the internal powering sequence and delays, especially if the IC has more than one power supply. Therefore, give special consideration to power coupling capacitance, power wiring, width of ground wiring, and routing of connections.

8. Operation Under Strong Electromagnetic Field

Operating the IC in the presence of a strong electromagnetic field may cause the IC to malfunction.

9. Testing on Application Boards

When testing the IC on an application board, connecting a capacitor directly to a low-impedance output pin may subject the IC to stress. Always discharge capacitors completely after each process or step. The IC's power supply should always be turned off completely before connecting or removing it from the test setup during the inspection process. To prevent damage from static discharge, ground the IC during assembly and use similar precautions during transport and storage.

10. Inter-pin Short and Mounting Errors

Ensure that the direction and position are correct when mounting the IC on the PCB. Incorrect mounting may result in damaging the IC. Avoid nearby pins being shorted to each other especially to ground, power supply and output pin. Inter-pin shorts could be due to many reasons such as metal particles, water droplets (in very humid environment) and unintentional solder bridge deposited in between pins during assembly to name a few.

11. Unused Input Pins

Input pins of an IC are often connected to the gate of a MOS transistor. The gate has extremely high impedance and extremely low capacitance. If left unconnected, the electric field from the outside can easily charge it. The small charge acquired in this way is enough to produce a significant effect on the conduction through the transistor and cause unexpected operation of the IC. So unless otherwise specified, unused input pins should be connected to the power supply or ground line.

Operational Notes – continued

12. Regarding the Input Pin of the IC

This monolithic IC contains P+ isolation and P substrate layers between adjacent elements in order to keep them isolated. P-N junctions are formed at the intersection of the P layers with the N layers of other elements, creating a parasitic diode or transistor. For example (refer to figure below):

When GND > Pin A and GND > Pin B, the P-N junction operates as a parasitic diode. When GND > Pin B, the P-N junction operates as a parasitic transistor.

Parasitic diodes inevitably occur in the structure of the IC. The operation of parasitic diodes can result in mutual interference among circuits, operational faults, or physical damage. Therefore, conditions that cause these diodes to operate, such as applying a voltage lower than the GND voltage to an input pin (and thus to the P substrate) should be avoided.

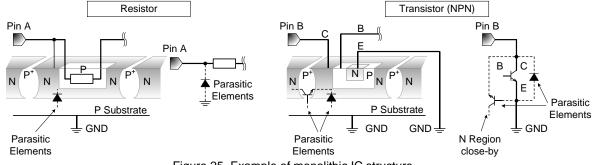
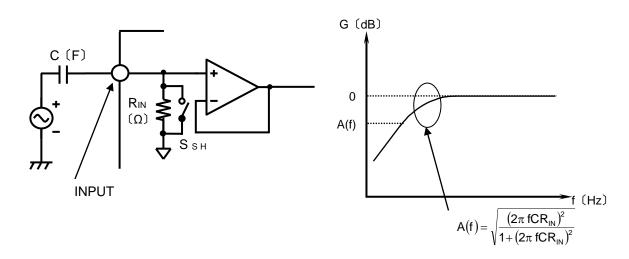


Figure 25. Example of monolithic IC structure

13. About a Signal Input Part

(a) About Input Coupling Capacitor Constant Value

The constant value of input coupling capacitor C(F) is decided with respect to the input impedance $R_{IN}(\Omega)$ at the input signal terminal of the IC that would be sufficient to form an RC characterized HPF.



(b) About the Input Selector SHORT

SHORT mode is the command which makes switch $S_{SH} = ON$ of input selector part so that the input impedance R_{IN} of all terminals becomes small. Switch S_{SH} is OFF when SHORT command is not selected. The constant time brought about by the small resistance inside and the capacitor outside the LSI becomes small when this command is used. The charge time of the capacitor becomes short. Since SHORT mode turns ON the switch of S_{SH} and makes it low impedance, please use it at no signal condition.

14. About Mute Terminal (Pin 15) when Power Supply is OFF

There should be no applied voltage to Mute terminal (Pin 15) when power-supply is OFF. If in case voltage is supplied to MUTE terminal, please insert a series resistor (about $2.2k\Omega$) to Mute terminal. (Please refer to Application Circuit Diagram.)

Operational Notes – continued

15. About Mixing

(a) <u>About Specification of Fader -∞ at Mixing ON.</u>

Mixed signal is added to the Main signal together with the Fader Gain (+15dB to -79dB) shown in the figure below. When Fader is set up in $-\infty$, the signal after MIX is added with MUTE because the $-\infty$ circuit of Fader is in the step after the addition circuit.

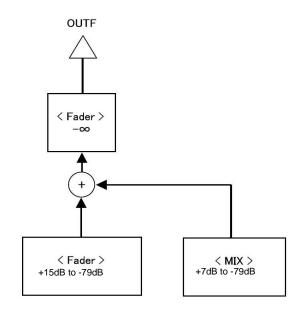
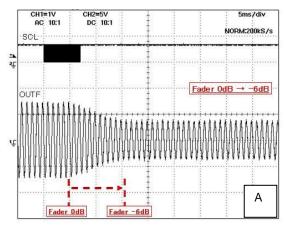


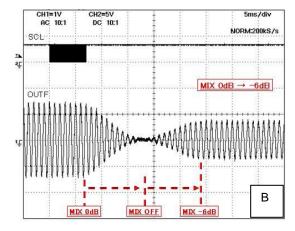
Figure 26. About Front Fader and Mixing

(b) About Advanced switching of Mixing Gain/ATT

When advanced switching of Mixing Gain/ATT works, Mixing becomes a switching movement that it passes through the state of Mixing OFF like what is shown in Figure B (from present setup of Mixing Gain/ATT to Mixing OFF to a target setup of Mixing Gain/ATT).



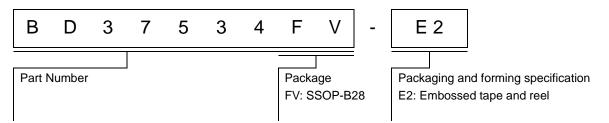
Fader Gain/ATT 0dB to -6dB Advanced Switching



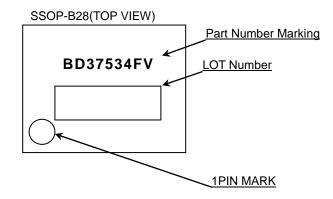
Mixing Gain/ATT 0dB to -6dB Advanced Switching

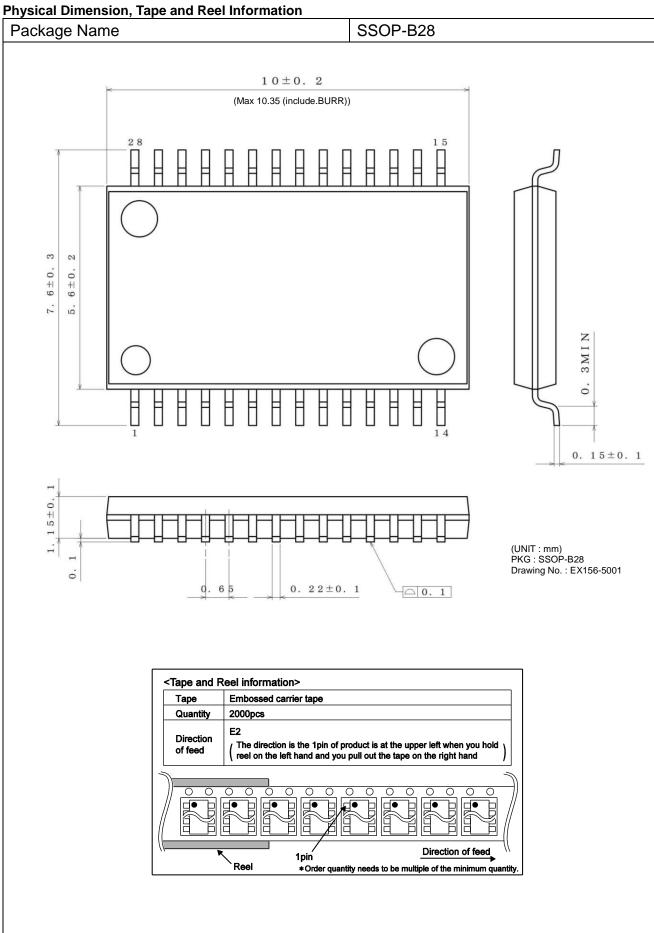
Figure 27. Advanced Switching Movement when Mixing Gain/ATT is changed

Ordering Information



Marking Diagram





Revision History

Date	Revision	Changes
16.Dec.2015	001	New Release

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JAPAN	USA	EU	CHINA
CLASSⅢ	CLASSⅢ	CLASS II b	
CLASSⅣ	CLASSII	CLASSⅢ	CLASSII

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 - [e] Use of our Products in proximity to heat-producing components, plastic cords, or other flammable items
 - [f] Sealing or coating our Products with resin or other coating materials
 - [g] Use of our Products without cleaning residue of flux (even if you use no-clean type fluxes, cleaning residue of flux is recommended); or Washing our Products by using water or water-soluble cleaning agents for cleaning residue after soldering
 - [h] Use of the Products in places subject to dew condensation
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For details, please refer to ROHM Mounting specification

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 - [b] the temperature or humidity exceeds those recommended by ROHM
 - [c] the Products are exposed to direct sunshine or condensation
 - [d] the Products are exposed to high Electrostatic
- 2. Even under ROHM recommended storage condition, solderability of products out of recommended storage time period may be degraded. It is strongly recommended to confirm solderability before using Products of which storage time is exceeding the recommended storage time period.
- 3. Store / transport cartons in the correct direction, which is indicated on a carton with a symbol. Otherwise bent leads may occur due to excessive stress applied when dropping of a carton.
- 4. Use Products within the specified time after opening a humidity barrier bag. Baking is required before using Products of which storage time is exceeding the recommended storage time period.

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